# TLE42694

Low Dropout Fixed Voltage Regulator

**Automotive Power** 





### Low Dropout Fixed Voltage Regulator

#### **TLE42694**





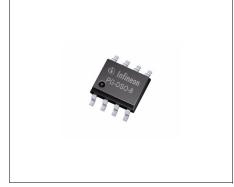
### 1 Overview

#### **Features**

- Output Voltage 5 V ± 2%
- Ouput Current up to 150 mA
- Very Low Current Consumption
- Early Warning
- Power-on and Undervoltage Reset with Programmable Delay Time
- Reset Low Down to V<sub>O</sub> = 1 V
- Adjustable Reset Threshold
- Very Low Dropout Voltage
- Output Current Limitation
- Reverse Polarity Protection
- · Overtemperature Protection
- · Suitable for Use in Automotive Electronics
- Wide Temperature Range from -40 °C up to 150 °C
- Input Voltage Range from -42 V to 45 V
- Integrated Pull-Up Resistors at Logic Outputs
- Green Product (RoHS compliant)
- AEC Qualified

#### **Description**

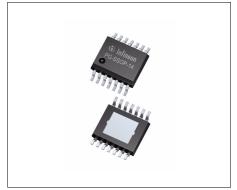
The TLE 42694 is a monolithic integrated low dropout voltage regulator, especially designed for automotive applications. An input voltage up to 45 V is regulated to an output voltage of 5.0 V. The component is able to drive loads up to 150 mA. It is short-circuit proof by the implemented output current limitation and has an integrated overtemperature shutdown. A reset signal is generated for an output voltage  $V_{\rm Q,rt}$  of typically 4.65 V. This threshold can be decreased by an external resistor divider. The power-on reset delay time can be programmed by the external delay capacitor. The additional sense comparator provides an early warning function: Any voltage (e.g. the input voltage) can be monitored, an under-voltage condition is indicated by setting the comparator's output to low. The reset and sense output are internally connected to the output Q via a pull-up resistor. If these



PG-DSO-8



PG-DSO-14



PG-SSOP-14 exposed pad

integrated resistors are not desired, the TLE42794 can be used instead of the TLE42694.

Туре	Package	Marking
TLE42694G	PG-DSO-8	42694G
TLE42694GM	PG-DSO-14	42694GM
TLE42694E	PG-SSOP-14 exposed pad	42694E

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Overview

#### **Dimensioning Information on External Components**

The input capacitor  $C_{l}$  is recommended for compensation of line influences. The output capacitor  $C_{Q}$  is necessary for the stability of the control loop.

#### **Circuit Description**

The control amplifier compares a reference voltage to a voltage that is proportional to the output voltage and drives the base of the series transistor via a buffer. Saturation control as a function of the load current prevents any oversaturation of the power element. The component also has a number of internal circuits for protection against:

- Overload
- Overtemperature
- Reverse polarity

**Block Diagram** 

# 2 Block Diagram

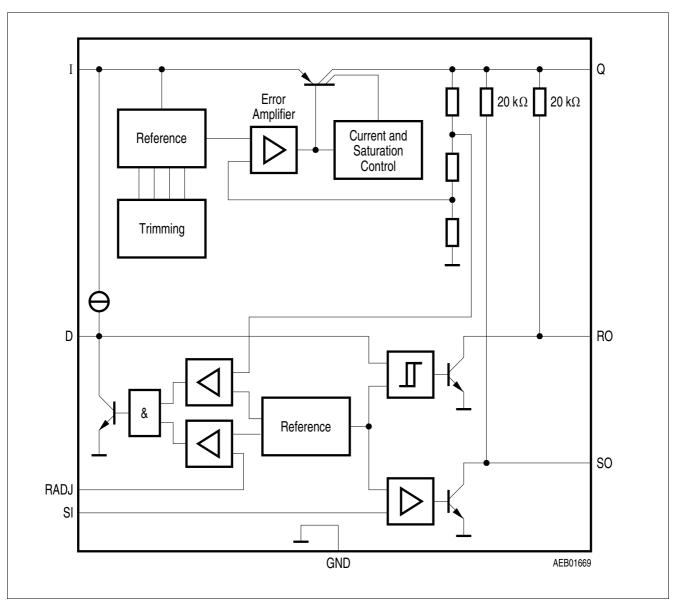


Figure 1 Block Diagram



**Pin Configuration** 

# 3 Pin Configuration

# 3.1 Pin Assignment PG-DSO-8

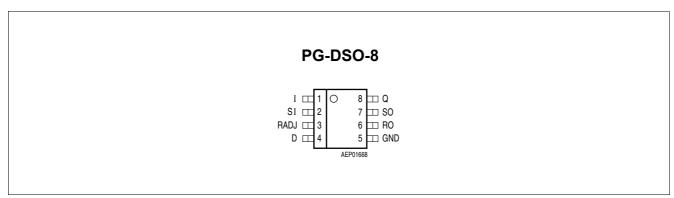


Figure 2 Pin Configuration (top view)

## 3.2 Pin Definitions and Functions PG-DSO-8

#### Table 1

Pin	Symbol	Function
1	I	Input for compensating line influences, a capacitor to GND close to the IC terminals is recommended
2	SI	Sense Input connect the voltage to be monitored; connect to Q if the sense comparator is not needed
3	RADJ	Reset Threshold Adjust connect an external voltage divider to adjust reset threshold; connect to GND for using internal threshold
4	D	Reset Delay Timing connect a ceramic capacitor to GND for adjusting the reset delay time; leave open if the reset function is not needed
5	GND	Ground
6	RO	Reset Output open collector output; internally linked to the output via a $20k\Omega$ pull-up resistor; leave open if the reset function is not needed
7	SO	Sense Output open collector output; internally linked to the output via a $20k\Omega$ pull-up resistor; leave open if the sense comparator is not needed
8	Q	Output block to GND with a capacitor close to the IC terminals, respecting the values give for its capacitance $C_Q$ and ESR in "Functional Range" on Page 9



**Pin Configuration** 

# 3.3 Pin Assignment PG-DSO-14

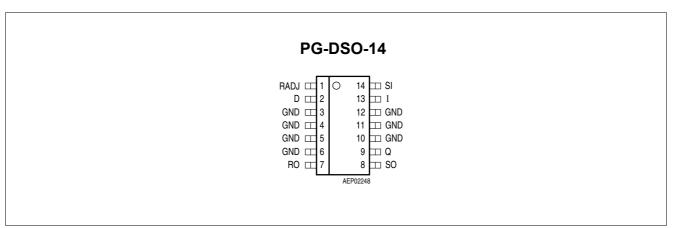


Figure 3 Pin Configuration (top view)

### 3.4 Pin Definitions and Functions PG-DSO-14

#### Table 2

Pin	Symbol	Function
1	RADJ	Reset Threshold Adjust
		connect an external voltage divider to adjust reset threshold;
		connect to GND for using internal threshold
2	D	Reset Delay Timing
		connect a ceramic capacitor to GND for adjusting the reset delay time;
		leave open if the reset function is not needed
3, 4, 5, 6	GND	Ground
		all pins must be connected to GND
7	RO	Reset Output
		open collector output; internally linked to the output via a $20k\Omega$ pull-up resistor;
		leave open if the reset function is not needed
8	SO	Sense Output
		open collector output; internally linked to the output via a $20k\Omega$ pull-up resistor;
		leave open if the sense comparator is not needed
9	Q	Output
		block to GND with a capacitor close to the IC terminals, respecting the values given
		for its capacitance $C_Q$ and ESR in the table "Functional Range" on Page 9
10, 11, 12	GND	Ground
		all pins must be connected to GND
13	I	Input
		for compensating line influences, a capacitor to GND close to the IC terminals is
		recommended
14	SI	Sense Input
		connect the voltage to be monitored;
		connect to Q if the sense comparator is not needed



**Pin Configuration** 

# 3.5 Pin Assignment PG-SSOP-14 exposed pad

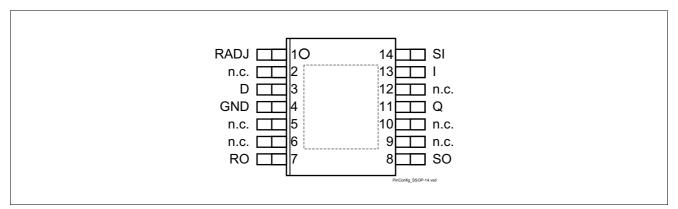


Figure 4 Pin Configuration (top view)

# 3.6 Pin Definitions and Functions PG-SSOP-14 exposed pad

#### Table 3

Pin	Symbol	Function
1	RADJ	Reset Threshold Adjust
		connect an external voltage divider to adjust reset threshold;
		connect to GND for using internal threshold
2, 5, 6	n.c.	not connected
3	D	Reset Delay Timing
		connect a ceramic capacitor to GND for adjusting the reset delay time;
		leave open if the reset function is not needed
4	GND	Ground
		all pins must be connected to GND
7	RO	Reset Output
		open collector output; internally linked to the output via a $20k\Omega$ pull-up resistor;
		leave open if the reset function is not needed
8	SO	Sense Output
		open collector output; internally linked to the output via a $20k\Omega$ pull-up resistor;
		leave open if the sense comparator is not needed
9, 10, 12	n.c.	not connected
11	Q	Output
		block to GND with a capacitor close to the IC terminals, respecting the values given
		for its capacitance $C_Q$ and ESR in the table "Functional Range" on Page 7
13	I	Input
		for compensating line influences, a capacitor to GND close to the IC terminals is
		recommended
14	SI	Sense Input
		connect the voltage to be monitored;
		connect to Q if the sense comparator is not needed
Pad	-	Exposed Pad
		connect to heatsink area;
		connect to GND



**General Product Characteristics** 

### 4 General Product Characteristics

## 4.1 Absolute Maximum Ratings

## Absolute Maximum Ratings 1)

-40 °C  $\leq$  T  $_{j}$   $\leq$  150 °C; all voltages with respect to ground, positive current flowing into pin (unless otherwise specified)

Pos.	Parameter	Symbol	Lin	nit Values	Unit	Conditions
			Min.	Max.		
Input, S	Sense Input		<del>-</del>	*		
4.1.1	Voltage	$V_{I},V_{SI}$	-40	45	V	_
Output	, Reset Output, Sense Output, Reset	Delay	-	<u>'</u>		,
4.1.2	Voltage	$V_{\mathrm{Q}},V_{\mathrm{RO}},\ V_{\mathrm{SO}},V_{\mathrm{D}}$	-0.3	7	V	_
Reset 1	Threshold	,				'
4.1.3	Voltage	$V_{RADJ}$	-0.3	7	V	_
4.1.4	Current	$I_{RADJ}$	-10	10	mA	_
Tempe	rature	·				
4.1.5	Junction Temperature	$T_{j}$	-40	150	°C	_
4.1.6	Storage Temperature	$T_{ m stg}$	-50	150	°C	_
ESD St	sceptibility					•
4.1.7	Human Body Model (HBM) <sup>2)</sup>	Voltage	-2	2	kV	_
4.1.8	Charged Device Model (CDM) <sup>3)</sup>	Voltage	-1	1	kV	_

<sup>1)</sup> not subject to production test, specified by design

Note: Maximum ratings are absolute ratings; exceeding any one of these values may cause irreversible damage to the integrated circuit. Integrated protection functions are designed to prevent IC destruction under fault conditions. Fault conditions are considered as outside normal operating range. Protections functions are not designed for continuous repetitive operation.

<sup>2)</sup> ESD HBM Test according to AEC-Q100-002 - JESD22-A114

<sup>3)</sup> ESD CDM Test according to ESDA ESD-STM5.3.1



#### **General Product Characteristics**

# 4.2 Functional Range

Pos.	Parameter	Symbol	Lin	nit Values	Unit	Conditions
			Min.	Max.		
4.2.1	Input Voltage	$V_1$	5.5	45	V	_
4.2.2	Output Capacitor's Requirements	$C_{Q}$	10	_	μF	_1)
	for Stability	$ESR(C_{Q})$	_	3	Ω	_2)
4.2.3	Junction Temperature	$T_{\rm j}$	-40	150	°C	_

<sup>1)</sup> the minimum output capacitance requirement is applicable for a worst case capacitance tolerance of 30%

Note: Within the functional range the IC operates as described in the circuit description. The electrical characteristics are specified within the conditions given in the related electrical characteristics table.

<sup>2)</sup> relevant ESR value at f = 10 kHz



#### **General Product Characteristics**

#### 4.3 Thermal Resistance

Pos.	Parameter	Symbol	L	imit Va	lue	Unit	Conditions	
			Min.	Тур.	Max.			
TLE426	694G (PG-DSO-8)							
4.3.4	Junction to Soldering Point <sup>1)</sup>	$R_{thJSP}$	_	80	_	K/W	measured to pin 5	
4.3.5	Junction to Ambient <sup>1)</sup>	$R_{thJA}$	_	113	_	K/W	2)	
4.3.6			_	170	_	K/W	Footprint only <sup>3)</sup>	
4.3.7			_	142	-	K/W	300mm <sup>2</sup> heatsink area on PCB <sup>3)</sup>	
4.3.8			_	136	_	K/W	600mm <sup>2</sup> heatsink area on PCB <sup>3)</sup>	
TLE426	694GM (PG-DSO-14)							
4.3.9	Junction to Soldering Point <sup>1)</sup>	$R_{\mathrm{thJSP}}$	_	27	_	K/W	measured to group of pins 3, 4, 5, 10, 11, 12	
4.3.10	Junction to Ambient <sup>1)</sup>	$R_{thJA}$	_	63	_	K/W	2)	
4.3.11			_	104	_	K/W	Footprint only <sup>3)</sup>	
4.3.12			_	73	_	K/W	300mm <sup>2</sup> heatsink area on PCB <sup>3)</sup>	
4.3.13			_	65	_	K/W	600mm <sup>2</sup> heatsink area on PCB <sup>3)</sup>	
TLE426	394E (PG-SSOP-14 exposed page	d)						
4.3.14	Junction to Soldering Point <sup>1)</sup>	$R_{thJSP}$	_	10	_	K/W	measured to pin 5	
4.3.15	Junction to Ambient <sup>1)</sup>	$R_{thJA}$	_	47	_		2)	
4.3.16			_	145	_	K/W	Footprint only <sup>3)</sup>	
4.3.17			_	63	_	K/W	300mm <sup>2</sup> heatsink area on PCB <sup>3)</sup>	
4.3.18			_	53	_	K/W	600mm <sup>2</sup> heatsink area on PCB <sup>3)</sup>	

<sup>1)</sup> not subject to production test, specified by design

<sup>2)</sup> Specified  $R_{\text{thJA}}$  value is according to Jedec JESD51-2,-5,-7 at natural convection on FR4 2s2p board; The Product (Chip+Package) was simulated on a 76.2 x 114.3 x 1.5 mm³ board with 2 inner copper layers (2 x 70µm Cu, 2 x 35µm Cu). Where applicable a thermal via array under the exposed pad contacted the first inner copper layer.

<sup>3)</sup> Specified  $R_{\text{thJA}}$  value is according to JEDEC JESD 51-3 at natural convection on FR4 1s0p board; The Product (Chip+Package) was simulated on a 76.2 × 114.3 × 1.5 mm<sup>3</sup> board with 1 copper layer (1 x 70µm Cu).



# 5 Block Description and Electrical Characteristics

### 5.1 Voltage Regulator

The output voltage  $V_{\rm Q}$  is controlled by comparing a portion of it to an internal reference and driving a PNP pass transistor accordingly. The control loop stability depends on the output capacitor  $C_{\rm Q}$ , the load current, the chip temperature and the poles/zeros introduced by the integrated circuit. To ensure stable operation, the output capacitor's capacitance and its equivalent series resistor ESR requirements given in the table "Functional Range" on Page 9 have to be maintained. For details see also the typical performance graph "Output Capacitor Series Resistor ESR( $C_{\rm Q}$ ) versus Output Current  $C_{\rm Q}$ " on Page 14. As the output capacitor also has to buffer load steps it should be sized according to the application's needs.

An input capacitor  $C_l$  is strongly recommended to compensate line influences. Connect the capacitors close to the component's terminals.

A protection circuitry prevents the IC as well as the application from destruction in case of catastrophic events. These safeguards contain an output current limitation, a reverse polarity protection as well as a thermal shutdown in case of overtemperature.

In order to avoid excessive power dissipation that could never be handled by the pass element and the package, the maximum output current is decreased at input voltages above  $V_1$  = 22 V.

The thermal shutdown circuit prevents the IC from immediate destruction under fault conditions (e.g. output continuously short-circuited) by switching off the power stage. After the chip has cooled down, the regulator restarts. This leads to an oscillatory behaviour of the output voltage until the fault is removed. However, junction temperatures above 150 °C are outside the maximum ratings and therefore significantly reduce the IC's lifetime.

The TLE 42694 allows a negative supply voltage. In this fault condition, small currents are flowing into the IC, increasing its junction temperature. This has to be considered for the thermal design, respecting that the thermal protection circuit is not operating during reverse polarity conditions.

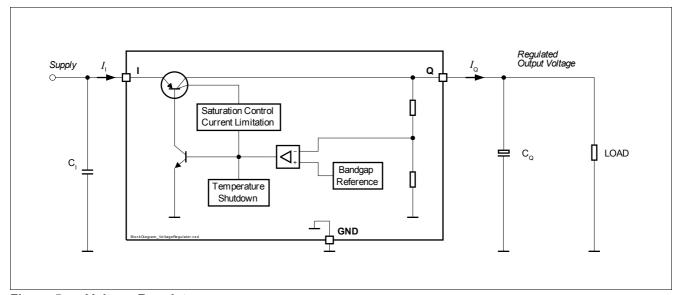


Figure 5 Voltage Regulator



## **Electrical Characteristics Voltage Regulator**

 $V_{\rm I}$  = 13.5 V, -40 °C  $\leq$   $T_{\rm j} \leq$ 150 °C, all voltages with respect to ground, positive current flowing into pin (unless otherwise specified)

Pos.	Parameter	Symbol		Limit Val	ues	Unit	Conditions
			Min.	Тур.	Max.		
5.1.1	Output Voltage	$V_{Q}$	4.9	5.0	5.1	V	100 $\mu$ A < $I_Q$ < 100 mA 6 V < $V_I$ < 18 V
5.1.2	Output Current Limitation	$I_{\mathrm{Q,max}}$	150	200	500	mA	$V_{\rm Q}$ = 4.8V
5.1.3	Load Regulation steady-state	$\Delta V_{ m Q,load}$	-30	-15	-	mV	$I_{\rm Q}$ = 5 mA to 100 mA $V_{\rm I}$ = 6 V
5.1.4	Line Regulation steady-state	$\Delta V_{ m Q,line}$	_	10	40	mV	$V_{\rm I}$ = 6 V to 32 V $I_{\rm Q}$ = 5 mA
5.1.5	Dropout Voltage <sup>1)</sup> $V_{dr} = V_{l} - V_{Q}$	$V_{dr}$	_	250	500	mV	$I_{\rm Q}$ = 100 mA
5.1.6	Overtemperature Shutdown Threshold	$T_{ m j,sd}$	151	-	200	°C	$T_{\rm j}$ increasing <sup>2)</sup>
5.1.7	Overtemperature Shutdown Threshold Hysteresis	$T_{ m j,sdh}$	-	15	_	°C	$T_{\rm j}$ decreasing <sup>2)</sup>
5.1.8	Power Supply Ripple Rejection <sup>2)</sup>	PSRR	_	70	-	dB	$f_{\text{ripple}}$ = 100 Hz $V_{\text{ripple}}$ = 0.5 Vp

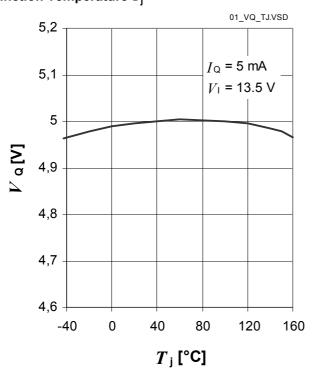
<sup>1)</sup> measured when the output voltage  $V_{\rm Q}$  has dropped 100mV from the nominal value obtained at  $V_{\rm I}$  = 13.5V

<sup>2)</sup> not subject to production test, specified by design

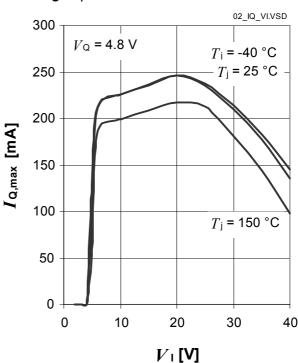


#### **Typical Performance Characteristics Voltage Regulator**

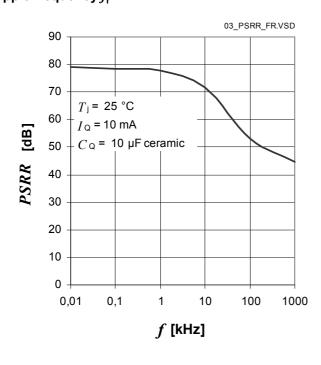
# Output Voltage $V_{\rm Q}$ versus Junction Temperature $T_{\rm i}$



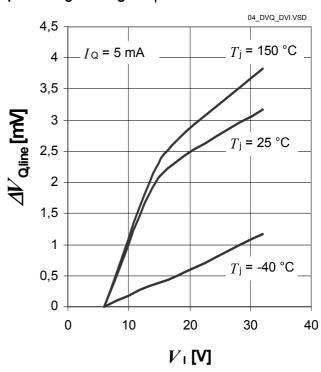
# Output Current $I_{\mathsf{Q}}$ versus Input Voltage $V_{\mathsf{I}}$



# Power Supply Ripple Rejection PSRR versus ripple frequency $f_r$



# Line Regulation $\Delta V_{\mathrm{Q,line}}$ versus Input Voltage Change $\Delta V_{\mathrm{I}}$



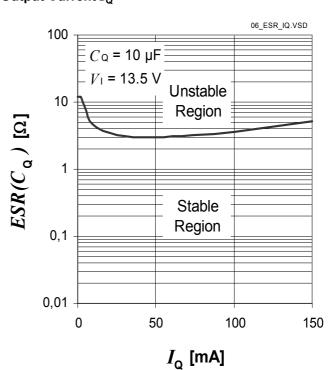


#### **Typical Performance Characteristics Voltage Regulator**

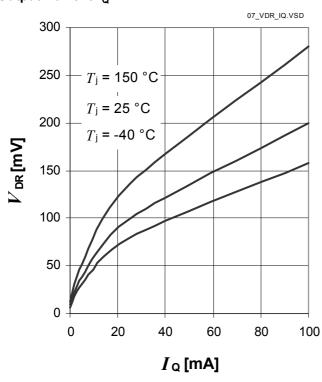
# Load Regulation $\Delta V_{ m Q,load}$ versus Output Current Change $\Delta I_{ m Q}$

# 05\_DVQ\_DIQ.VSD $V_1 = 13.5 \text{ V}$ -2 $\Delta V_{ m Q\,load}$ [mV] -6 $T_{i} = -40 \, ^{\circ}\text{C}$ $T_{i} = 25 \, ^{\circ}\text{C}$ $T_{\rm j}$ = 150 °C -10 -12 -14 0 40 20 60 80 100 $I_{\rm Q}$ [mA]

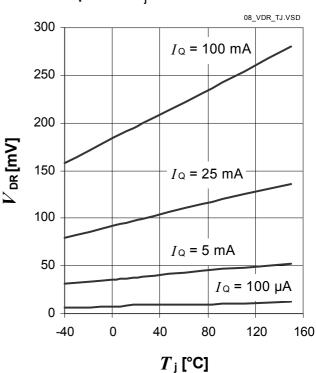
# Output Capacitor Series Resistor $ESR(C_{\rm Q})$ versus Output Current $I_{\rm Q}$



# Dropout Voltage $V_{\rm dr}$ versus Output Current $I_{\rm Q}$



# Dropout Voltage $V_{\rm dr}$ versus Junction Temperature $T_{\rm i}$





# 5.2 Current Consumption

#### **Electrical Characteristics Current Consumption**

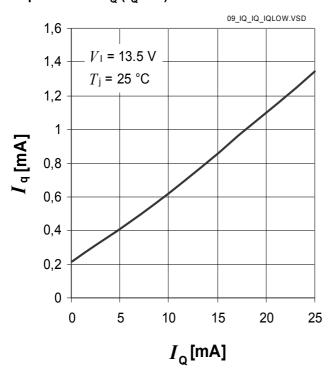
 $V_{\rm I}$  = 13.5 V, -40 °C  $\leq$   $T_{\rm j}$   $\leq$  150 °C, positive current flowing into pin (unless otherwise specified)

Pos.	Parameter Symbol Limit Values				lues	Unit	Conditions
			Min.	Тур.	Max.		
5.2.1	Current Consumption	$I_{q}$	_	210	280	μΑ	$I_{\rm Q}$ = 100 $\mu {\rm A}$
	$I_{q} = I_{l} - I_{Q}$	1					$T_{\rm i}$ = 25 °C
5.2.2			_	240	300	μΑ	$I_{\rm Q}$ = 100 $\mu$ A
							$T_{\rm i} \leq 85~{\rm ^{\circ}C}$
5.2.3			_	0.7	1	mA	$I_{\rm Q}$ = 10 mA
5.2.4			_	3.5	8	mA	$I_{\rm Q}$ = 50 mA

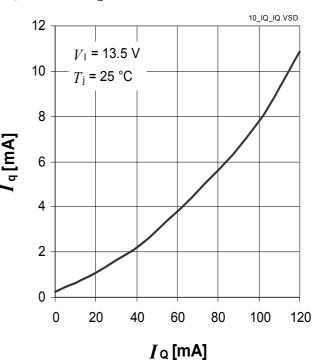


### **Typical Performance Characteristics Current Consumption**

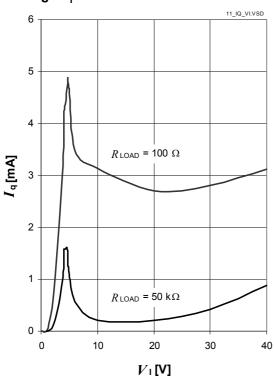
# Current Consumption $I_{\rm q}$ versus Output Current $I_{\rm Q}$ ( $I_{\rm Q}$ low)



# Current Consumption $I_{\rm q}$ versus Output Current $I_{\rm Q}$



# Current Consumption $I_{\rm q}$ versus Input Voltage $V_{\rm l}$





#### 5.3 Reset Function

The reset function provides several features:

#### **Output Undervoltage Reset:**

An output undervoltage condition is indicated by setting the Reset Output RO to "low". This signal might be used to reset a microcontroller during low supply voltage.

#### **Power-On Reset Delay Time:**

The power-on reset delay time  $t_{\rm rd}$  allows a microcontoller and oscillator to start up. This delay time is the time frame from exceeding the reset switching threshold  $V_{\rm RT}$  until the reset is released by switching the reset output "RO" from "low" to "high". The power-on reset delay time  $t_{\rm rd}$  is defined by an external delay capacitor  $C_{\rm D}$  connected to pin D charged by the delay capacitor charge current  $I_{\rm D,ch}$  starting from  $V_{\rm D}$  = 0 V.

If the application needs a power-on reset delay time  $t_{rd}$  different from the value given in **Item 5.3.8**, the delay capacitor's value can be derived from the specified values in **Item 5.3.8** and the desired power-on delay time:

$$C_{D} = \frac{t_{rd, new}}{t_{rd}} \times 100 nF$$

with

- C<sub>D</sub>: capacitance of the delay capacitor to be chosen
- t<sub>rd,new</sub>: desired power-on reset delay time
- t<sub>rd</sub>: power-on reset delay time specified in this datasheet

For a precise calculation also take the delay capacitor's tolerance into consideration.

#### **Reset Reaction Time:**

The reset reaction time avoids that short undervoltage spikes trigger an unwanted reset "low" signal. The reset reaction rime  $t_{\rm rr}$  considers the internal reaction time  $t_{\rm rr,int}$  and the discharge time  $t_{\rm rr,d}$  defined by the external delay capacitor  $C_{\rm D}$  (see typical performance graph for details). Hence, the total reset reaction time becomes:

$$t_{rr} = t_{rd, int} + t_{rr, d}$$

with

- t<sub>rr</sub>: reset reaction time
- t<sub>rr.int</sub>: internal reset reaction time
- $t_{rr,d}$ : reset discharge

#### Optional Reset Output Pull-Up Resistor $R_{RO.ext}$ :

The Reset Output RO is an open collector output with an integrated pull-up resistor. To improve the EMC behaviour of the component, an external pull-up resistor to the output  $V_{\rm Q}$  can be added. In **Table "Electrical Characteristics Reset Function" on Page 21** a minimum value for the external resistor  $R_{\rm RO\,ext}$  is given.



#### **Reset Adjust Function**

The undervoltage reset switching threshold can be adjusted according to the application's needs by connecting an external voltage divider ( $R_{\rm ADJ1}$ ,  $R_{\rm ADJ2}$ ) at pin RADJ. For selecting the default threshold connect pin RADJ to GND.

When dimensioning the voltage divider, take into consideration that there will be an additional current constantly flowing through the resistors.

With a voltage divider connected, the reset switching threshold  $V_{\mathrm{RT,new}}$  is calculated as follows:

$$V_{RT,\,new} = \frac{R_{ADJ,\,1} + R_{ADJ,\,2}}{R_{ADJ,\,2}} \times V_{RADJ,\,th}$$

#### with

- V<sub>RT.new</sub>: the desired new reset switching threshold
- $R_{ADJ1}$ ,  $R_{ADJ2}$ : resistors of the external voltage divider
- $V_{\rm RADJ,th}$ : reset adjust switching threshold given in Table "Electrical Characteristics Reset Function" on Page 21

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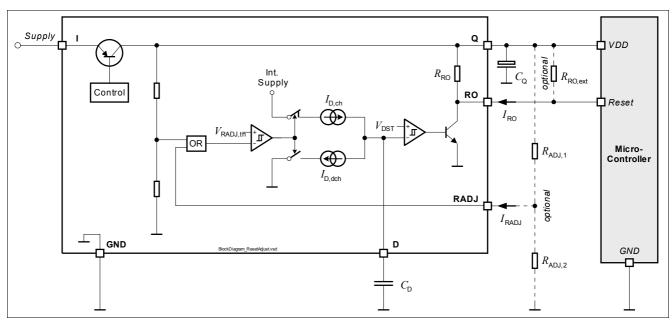


Figure 6 Block Diagram Reset Function



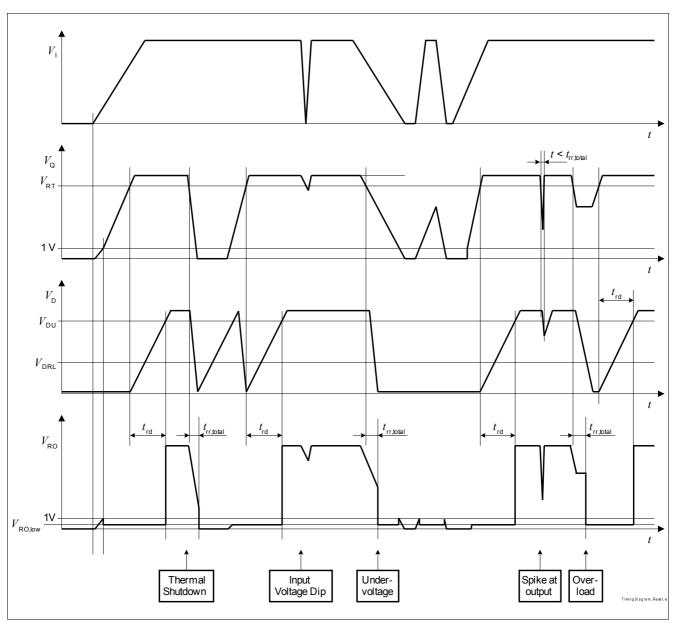


Figure 7 Timing Diagram Reset



#### **Electrical Characteristics Reset Function**

 $V_{\rm I}$  = 13.5 V, -40 °C  $\leq$   $T_{\rm j} \leq$  150 °C, all voltages with respect to ground, positive current flowing into pin (unless otherwise specified)

Pos.	Parameter	Symbol	Limit Values			Unit	Conditions
			Min.	Тур.	Max.		
Output	Undervoltage Reset						
5.3.1	Default Output Undervoltage Reset Switching Thresholds	$V_{RT}$	4.5	4.65	4.8	V	$V_{\mathrm{Q}}$ decreasing
Output	Undervoltage Reset Threshold Ad	justment					
5.3.2	Reset Adjust Switching Threshold	$V_{RADJ,th}$	1.26	1.35	1.44	V	$3.5 \text{ V} \le V_{Q} < 5 \text{ V}$
5.3.3	Reset Adjustment Range <sup>1)</sup>	$V_{\mathrm{RT,range}}$	3.50	_	4.65	V	_
Reset (	Output RO		1				
5.3.4	Reset Output Low Voltage	$V_{RO,low}$	_	0.1	0.4	V	1 V $\leq V_{\rm Q} \leq V_{\rm RT}$ no external $R_{\rm RO,ex}$
5.3.5	Reset Output Internal Pull-Up Resistor to $V_{\rm Q}$	$R_{RO}$	10	20	40	kΩ	_
5.3.6	Optional Reset Output External Pull-up Resistor to $V_{\rm Q}$	$R_{RO,ext}$	20	_	_	kΩ	$\begin{array}{c} 1~\mathrm{V} \leq V_\mathrm{Q} \leq V_\mathrm{RT};\\ V_\mathrm{RO} \leq 0.4~\mathrm{V} \end{array}$
Reset I	Delay Timing						
5.3.7	Delay Pin Output Voltage	$V_{D}$	-	_	5	V	_
5.3.8	Power On Reset Delay Time	$t_{\rm rd}$	17	28	39	ms	$C_{\rm D}$ = 100 nF
5.3.9	Upper Delay Switching Threshold	$V_{DU}$	_	1.8	_	V	-
5.3.10	Lower Delay Switching Threshold	$V_{DL}$	_	0.45	_	V	_
5.3.11	Delay Capacitor Charge Current	$I_{D,ch}$	_	6.5	_	μA	<i>V</i> <sub>D</sub> = 1 V
5.3.12	Delay Capacitor Reset Discharge Current	$I_{D,dch}$	_	70	_	mA	<i>V</i> <sub>D</sub> = 1 V
5.3.13	Delay Capacitor Discharge Time	$t_{\rm rr,d}$	_	1.9	3	μs	Calculated Value $t_{\rm rr,d} = C_{\rm D} * (V_{\rm DU} - V_{\rm DL}) / I_{\rm D,dch}$ $C_{\rm D} = 100~{\rm nF}$
5.3.14	Internal Reset Reaction Time	$t_{\rm rr,int}$	-	3	7	μs	$C_{\rm D}$ = 0 nF <sup>2)</sup>
5.3.15	Reset Reaction Time	$t_{ m rr,total}$	_	4.9	10	μs	Calculated Value $t_{\text{rr,total}} = t_{\text{rr,int}} + t_{\text{rr,d}}$ $C_{\text{D}} = 100 \text{ nF}$

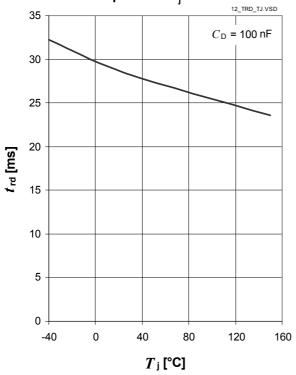
<sup>1)</sup>  $V_{\mathrm{RT}}$  is scaled linearly, in case the Reset Switching Threshold is modified

<sup>2)</sup> parameter not subject to production test; specified by design

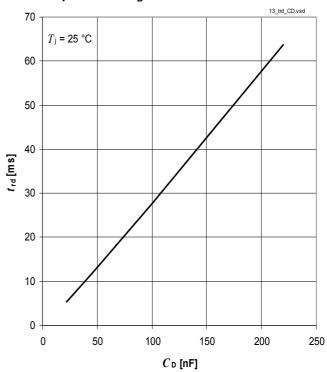


### **Typical Performance Characteristics**

# Power On Reset Delay Time $t_{\rm rd}$ versus Junction Temperature $T_{\rm i}$



# Power On Reset Delay Time $t_{\rm rd}$ versus Capacitance $C_{\rm D}$





## 5.4 Early Warning Function

The additional sense comparator provides an early warning function: Any voltage (e.g. the input voltage) can be monitored, an undervoltage condition is indicated by setting the comparator's output to low.

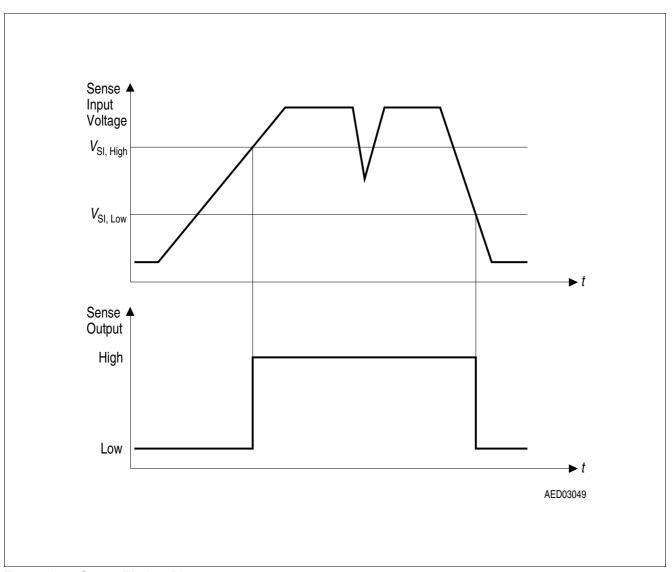


Figure 8 Sense Timing Diagram

## **Electrical Characteristics Early Warning Function**

 $V_{\rm I}$  = 13.5 V, -40 °C  $\leq T_{\rm j} \leq$  150 °C, all voltages with respect to ground, positive current flowing into pin (unless otherwise specified)

Pos.	Parameter	Symbol	I	Limit Val	ues	Unit	Conditions
			Min.	Тур.	Max.		
Sense	Comparator Input	<u> </u>					1
5.4.1	Sense Threshold High	$V_{\rm SI,high}$	1.24	1.31	1.38	V	_
5.4.2	Sense Threshold Low	$V_{SI,low}$	1.16	1.22	1.28	V	_
5.4.3	Sense Switching Hysteresis	$V_{\mathrm{SI,hy}}$	20	90	160	mV	_
5.4.4	Sense Input Current	$I_{SI}$	-1	-0.1	1	μΑ	_



## **Electrical Characteristics Early Warning Function**

 $V_{\rm I}$  = 13.5 V, -40 °C  $\leq$   $T_{\rm j} \leq$  150 °C, all voltages with respect to ground, positive current flowing into pin (unless otherwise specified)

Pos.	Parameter	<b>Symbol</b>		Limit Val	ues	Unit	Conditions
			Min.	Тур.	Max.		
Sense	Comparator Output		- 1				
5.4.5	Sense Output Low Voltage	$V_{SO,low}$	-	0.1	0.4	V	$V_{\rm SI} < V_{\rm SI,low}$ $V_{\rm I} > 5.5 \ {\rm V}$ no external $R_{\rm SO,e}$
5.4.6	Sense Output Internal Pull-Up Resistor to $V_{\rm Q}$	$R_{SO,int}$	10	20	40	kΩ	_
5.4.7	Optional Sense Output External Pull-up Resistor to $V_{\rm Q}$	$R_{SO,ext}$	20	_	_	kΩ	$V_{\rm I} > 5.5  {\rm V}$ $V_{\rm SO} \le 0.4  {\rm V}$

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**Package Outlines** 

# 6 Package Outlines

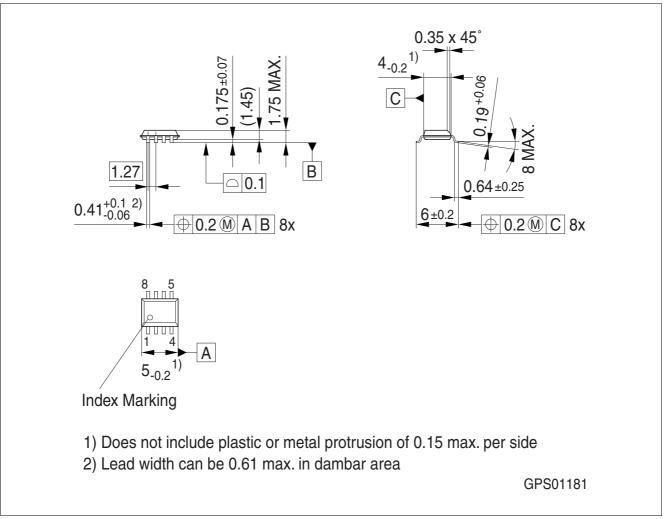


Figure 9 PG-DSO-8



### **Package Outlines**

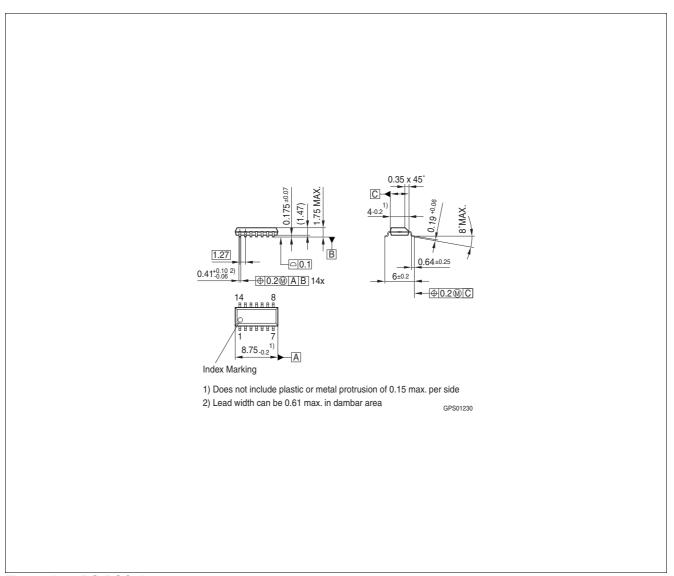


Figure 10 PG-DSO-14



**Package Outlines** 

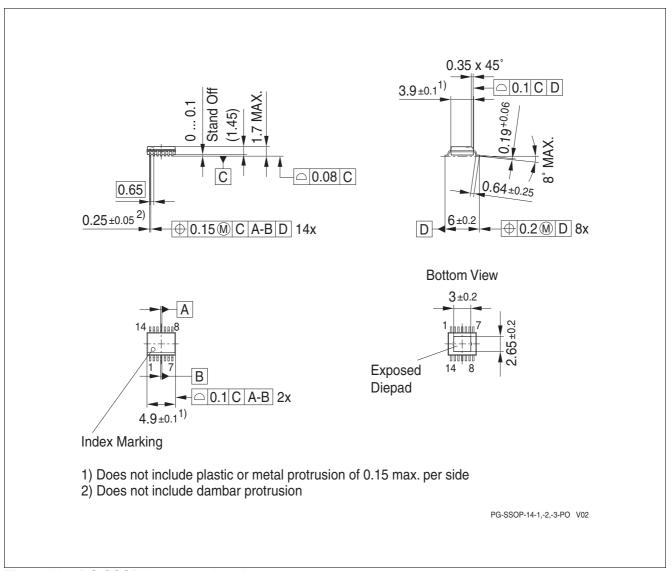


Figure 11 PG-SSOP-14 exposed pad

## **Green Product (RoHS compliant)**

To meet the world-wide customer requirements for environmentally friendly products and to be compliant with government regulations the device is available as a green product. Green products are RoHS-Compliant (i.e Pb-free finish on leads and suitable for Pb-free soldering according to IPC/JEDEC J-STD-020).



**Revision History** 

# 7 Revision History

Revision	Date	Changes
1.2	2010-11-26	Updated PG-SSOP-14 Exposed Pad connection
1.1	2008-10-07	package version TLE42694E in PG-SSOP-14 exposed pad and all related information added
		In "Overview" on Page 2 package graphic for PG-SSOP-14 exposed pad and product name "TLE42694E" added
		In Chapter 3 "Pin Assignment PG-SSOP-14 exposed pad" on Page 7 and "Pin Definitions and Functions PG-SSOP-14 exposed pad" on Page 7 added
		In "Thermal Resistance" on Page 10 values for TLE42694E added
		In "Package Outlines" on Page 25 outlines for TLE4269E added
1.0	2008-08-25	initial version data sheet

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